LM49270 Boomer® Audio Power Amplifier Series Filterless 2.2W Stereo Class D Audio Subsystem with OCL Headphone Amplifier, 3D Enhancement, and Headphone Sense

Check for Samples: LM49270

FEATURES

- **Stereo Filterless Class D Amplifier**
- **Selectable OCL/CC Headphone Amplifier**
- **Headphone Sense Ability**
- **TI's 3D Enhancement**
- **RF Suppression**
- **I^2C Control Interface**
- **32-Step Digital Volume Control**
- **6 Operating Modes**
- **Output Short Circuit Protection and Thermal Shutdown Protection**
- **Minimum External Components**
- **Click and Pop Suppression**
- **Micro-Power Shutdown**
- **Independent Speaker and Headphone Volume Controls**
- **Available in Space-Saving 28 Pin WQFN Package**

APPLICATIONS

- **Portable DVD Players**
- **Smart Phones**
- **PDAs**
- **Laptops**

KEY SPECIFICATIONS

- **Stereo Class D Amplifier Efficiency:**
  - **V_{DD} = 3.3V, 450mW/Ch into 8Ω 84%**
  - **V_{DD} = 5V, 1W/Ch into 8Ω 84%**
- **Quiescent Power Supply Current, V_{DD} = 3.3V**
  - **Speaker Mode 5.5 mA**
  - **Headphone Mode (OCL) 4 mA**
- **Power Output/Channel, V_{DD} = 5V**
  - **Class D Speaker Amplifier:**
    - **R_L = 4Ω, THD+N = ≤ 10% 2.3 W**
    - **R_L = 8Ω, THD+N = ≤ 1% 106 W**
  - **Headphone Amplifier:**
    - **R_L = 16Ω, THD+N = ≤ 1% 155 mW**
    - **R_L = 32Ω, THD+N = ≤ 1% 90 mW**
- **Shutdown Current 0.02µA**

DESCRIPTION

The LM49270 is a fully integrated audio subsystem designed for stereo multimedia applications. The LM49270 combines a 2.2W stereo Class D amplifier with a 155mW stereo headphone amplifier, volume control, headphone sense, and TI's unique 3D sound enhancement into a single device. The LM49270 uses flexible I^2C control interface for multiple application requirements.

The filterless stereo class D amplifiers delivers 2.2W/channel into a 4Ω load with less than 10% THD+N with a 5V supply. The headphone amplifier features Output Capacitor-less (OCL) architecture that eliminates the output coupling capacitors required by traditional headphone amplifiers.

The IC features a headphone sense input (HPS) that automatically detects the presence of a headphone and configures the device accordingly. The LM49270 can automatically switch from OCL headphone output to a line driver output. If the VOC pin is pulled to GND, the VOC amplifier is disabled and the VOC pin is internally set to GND. This feature allows the LM49270 to be used as a line driver in OCL mode without a GND conflict on the headphone jack sleeve. Additionally, the headphone amplifier can be configured as capacitively coupled (CC).

The LM49270 features a 32 step volume control for the headphone and stereo outputs. The device mode select and volume are controlled through an I^2C compatible interface.

Output short circuit and thermal shutdown protection prevent the device from being damaged during fault conditions. Superior click and pop suppression eliminates audible transients on power-up/down and during shutdown. The LM49270 is available in a space saving 28-pin, 5x5mm WQFN package.

These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

Typical Application

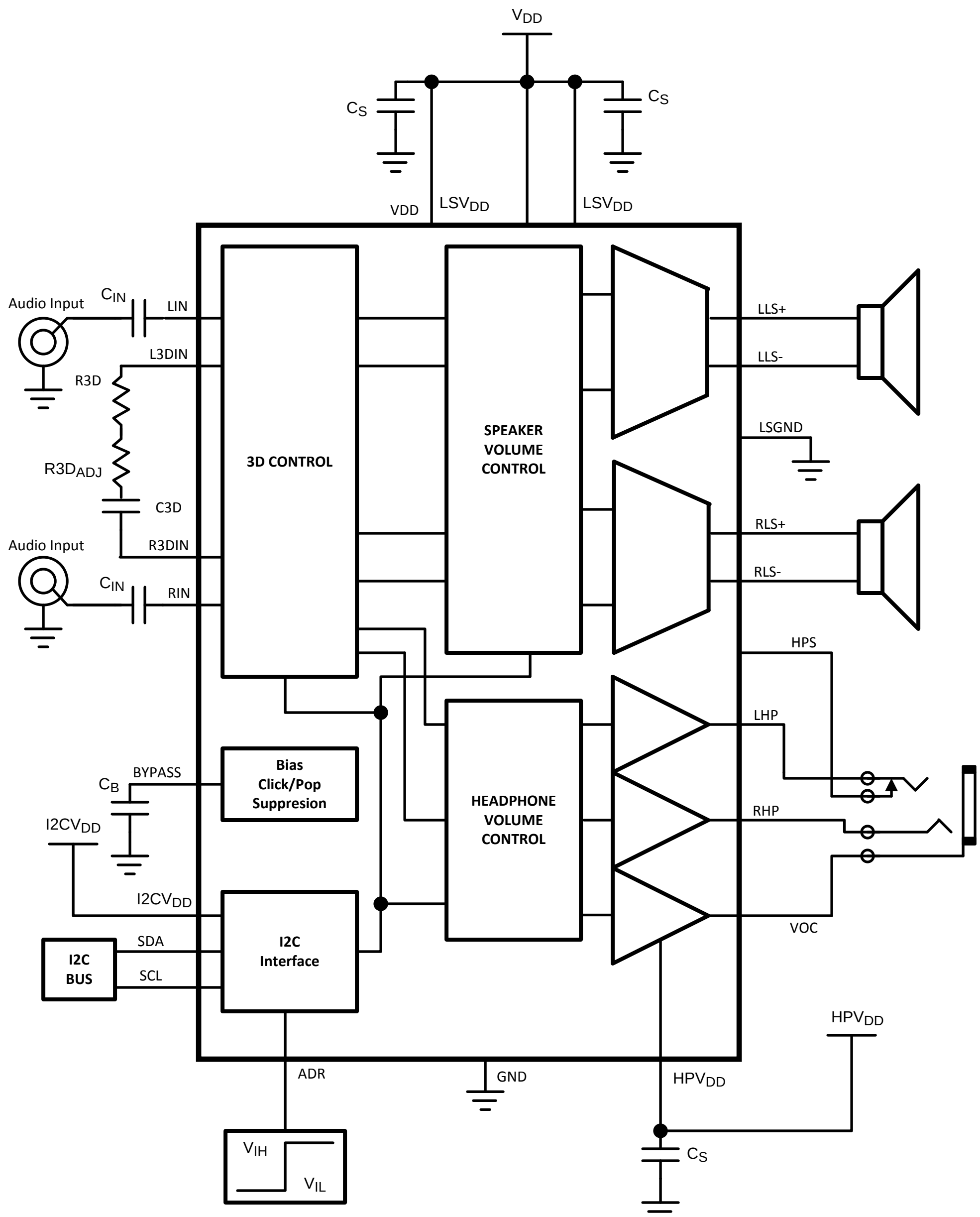

Figure 1. Typical Audio Amplifier Application Circuit

Connection Diagram

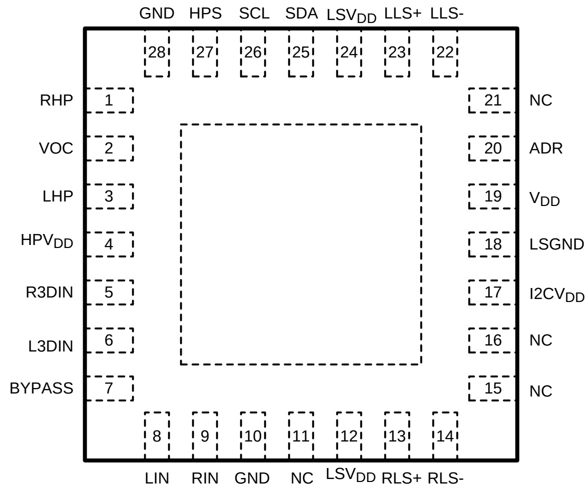

Figure 2. WQFN Package
5mm x 5mm x 0.8mm
Top View
See Package Number RSG0028A

Pin Descriptions

PIN	NAME	DESCRIPTION
1	RHP	Right channel headphone output
2	VOC	VDD/2 buffer output
3	LHP	Left channel headphone output
4	HPV_{DD}	Headphone supply input
5	R3DIN	Right channel 3D input
6	L3DIN	Left channel 3D input
7	BYPASS	Bias bypass
8	LIN	Left channel input
9	RIN	Right channel input
10	GND	Analog ground
11	NC	No connect
12	LSV_{DD}	Speaker supply voltage input
13	RLS+	Right channel non-inverting speaker output
14	RLS-	Right channel inverting speaker output
15	NC	No connect
16	NC	No connect
17	$I2CV_{DD}$	I2C supply voltage input
18	LSGND	Speaker ground
19	V_{DD}	Power supply
20	ADR	Address
21	NC	No connect
22	LLS-	Left channel inverting speaker output
23	LLS+	Left channel non-inverting speaker output
24	LSV_{DD}	Speaker supply voltage input
25	SDA	Serial data input
26	SCL	Serial clock input
27	HPS	Headphone sense input

Pin Descriptions (continued)

PIN	NAME	DESCRIPTION
28	GND	Headphone ground

Absolute Maximum Ratings [(1)(2)(3)]

Supply Voltage [(1)]		6.0V
Storage Temperature		−65°C to +150°C
Input Voltage		–0.3V to V_{DD} +0.3V
Power Dissipation [(4)]		Internally Limited
ESD Susceptibility [(5)]		2000V
ESD Susceptibility [(6)]		200V
Junction Temperature (T_{JMAX})		150°C
Thermal Resistance	θ_{JA}	35.1°C/W

(1) All voltages are measured with respect to the ground pin, unless otherwise specified.
(2) *Absolute Maximum Ratings* indicate limits beyond which damage to the device may occur. *Operating Ratings* indicate conditions for which the device is functional, but do not ensure specific performance limits. *Electrical Characteristics* state DC and AC electrical specifications under particular test conditions which specify performance limits. This assumes that the device is within the Operating Ratings. Specifications are not ensured for parameters where no limit is given, however, the typical value is a good indication of device performance.
(3) If Military/Aerospace specified devices are required, please contact the TI Sales Office/ Distributors for availability and specifications.
(4) The maximum power dissipation must be derated at elevated temperatures and is dictated by T_{JMAX}, θ_{JA}, and the ambient temperature, T_A. The maximum allowable power dissipation is $P_{DMAX} = (T_{JMAX} – T_A)/ \theta_{JA}$ or the number given in Absolute Maximum Ratings, whichever is lower. For the LM49270 see power derating currents for more information.
(5) Human body model, 100pF discharged through a 1.5kΩ resistor.
(6) Machine Model, 220pF–240pF discharged through all pins.

Operating Ratings [(1)]

Temperature Range	$T_{MIN} \leq T_A \leq T_{MAX}$	$-40°C \leq T_A \leq 85°C$
Supply Voltage (V_{DD}, LSV_{DD}, HPV_{DD})		$2.4V \leq V_{DD} \leq 5.5V$
I^2C Voltage (I^2CV_{DD})		$2.4V \leq I^2CV_{DD} \leq 5.5V$

(1) All voltages are measured with respect to the ground pin, unless otherwise specified.

Electrical Characteristics V_{DD} = 3.3V

[(1)]The following specifications apply for Headphone: A_V = 0dB, $R_{L(HP)}$ = 32Ω; for Loudspeakers: A_V = 6dB, $R_{L(SP)}$ = 15μH + 8Ω + 15μH , f = 1kHz, unless otherwise specified. Limits apply for T_A = 25°C.

Symbol	Parameter	Conditions	LM49270 Typical [(2)]	LM49270 Limit [(3) (4)]	Units (Limits)
I_{DD}	Supply Current	V_{IN} = 0, R_L = No Load, Both channels active Speaker ON, HP OFF Speaker OFF, CC HP ON Speaker OFF, OCL HP ON	 5.5 3 4	 7.6 4.7 5.75	 mA (max) mA (max) mA (max)
I_{SD}	Shutdown Supply Current		0.02	2	μA (max)
V_{OS}	Output Offset Voltage	Headphone Speaker	10 10	25 60	mV (max) mV (max)
P_{OUT}	Output Power	Speaker Mode, f = 1kHz			
		THD+N = 1% R_L = 4Ω R_L = 8Ω	 700 450	 400	 mW mW (min)
		THD+N = 10% R_L = 4Ω R_L = 8Ω	 870 560		 mW mW
		CC Headphone Mode, f = 1kHz			
		THD+N = 1% R_L = 16Ω R_L = 32Ω	 60 36	 30	 mW mW (min)
		THD+N = 10% R_L = 16Ω R_L = 32Ω	 74 55		 mW mW
		OCL Headphone Mode, f = 1kHz			
		THD+N = 1% R_L = 16Ω R_L = 32Ω	 60 36	 30	 mW mW (min)
		THD+N = 10% R_L = 16Ω R_L = 32Ω	 73 55		 mW mW
THD+N	Total Harmonic Distortion + Noise	Speaker Mode, f = 1kHz P_{OUT} = 100mW, R_L = 8Ω	0.02		%
		CC Headphone Mode, f = 1kHz P_{OUT} = 12mW, R_L = 32Ω	0.015		%
		OCL Headphone Mode, f = 1kHz P_{OUT} = 12mW, R_L = 32Ω	0.02		%
e_N	Noise	Speaker Mode, A-Wtg, Input Referred	47		μV
		CC Headphone Mode, A-Wtg, Input Referred	10		μV
		OCL Headphone Mode, A-Wtg, Input Referred	11		μV
η	Efficiency	Speaker Mode R_L = 8Ω	84		%
Xtalk	Crosstalk	Speaker Mode, f = 1kHz, V_{IN} = 1Vp-p	71		dB
		CC Headphone Mode, f = 1kHz, V_{IN} = 1Vp-p	70		dB
		OCL Headphone Mode, f = 1kHz, V_{IN} = 1Vp-p	55		dB

(1) All voltages are measured with respect to the ground pin, unless otherwise specified.
(2) Typicals are measured at 25°C and represent the parametric norm.
(3) Limits are specified to AOQL (Average Outgoing Quality Level).
(4) Data sheet min and max specification limits are specified by design, test, or statistical analysis.

Electrical Characteristics V_{DD} = 3.3V (continued)

[1] The following specifications apply for Headphone: A_V = 0dB, $R_{L(HP)}$ = 32Ω; for Loudspeakers: A_V = 6dB, $R_{L(SP)}$ = 15µH + 8Ω + 15µH , f = 1kHz, unless otherwise specified. Limits apply for T_A = 25°C.

Symbol	Parameter	Conditions	LM49270		Units (Limits)
			Typical [2]	Limit [3] [4]	
T_{ON}	Turn-on Time		30		ms
T_{OFF}	Turn-off Time		64		ms
Z_{IN}	Input Impedance	Maximum Gain	23.5		kΩ
		Minimum Gain	210		kΩ
A_V	Gain	Maximum Gain, Speaker Mode	30		dB
		Minimum Gain, Speaker Mode	–47		dB
		Maximum Gain, Headphone Mode	18		dB
		Minimum Gain, Headphone Mode	–59		dB
PSRR	Power Supply Rejection Ratio	Speaker Mode, V_{RIPPLE} = 200mVp-p Sine f = 217Hz f = 1kHz	68 68		dB dB
		Headphone Mode, V_{RIPPLE} = 200mVp-p Sine, CC Mode f = 217Hz f = 1kHz	73 73		dB dB
		Headphone Mode, V_{RIPPLE} = 200mVp-p Sine, OCL Mode f = 217Hz f = 1kHz	75 79		dB dB
$HPS_{(Th)}$	Headphone Sense Threshold	Detect Headphone		2.9	V (min)
		Detect no Headphone		1.8	V (max)

Electrical Characteristics V_{DD} = 5.0V

[1]The following specifications apply for Headphone" A_V = 0dB, $R_{L(HP)}$ = 32Ω,: for Loudspeakers: A_V = 6dB, $R_{L(SP)}$) = 15µH + 8Ω + 15µH, f = 1kHz unless otherwise specified. Limits apply for T_A = 25°C.

Symbol	Parameter	Conditions	LM49270		Units (Limits)
			Typical [2]	Limit [3] [4]	
I_{DD}	Supply Current	V_{IN} = 0, R_L = No Load, Both channels active Speaker ON, HP OFF Speaker OFF, CC HP ON Speaker OFF, OCL HP ON	8.5 3.6 4.7	12.4 5.5 6.5	mA (max) mA (max) mA (max)
I_{SD}	Shutdown Supply Current		0.15	2	µA (max)
V_{OS}	Output Offset Voltage	Headphone Speaker	10 10	25 60	mV (max) mV (max)

(1) All voltages are measured with respect to the ground pin, unless otherwise specified.
(2) Typicals are measured at 25°C and represent the parametric norm.
(3) Limits are specified to AOQL (Average Outgoing Quality Level).
(4) Data sheet min and max specification limits are specified by design, test, or statistical analysis.

Electrical Characteristics V_{DD} = 5.0V (continued)

(1) The following specifications apply for Headphone” A_V = 0dB, $R_{L(HP)}$ = 32Ω,: for Loudspeakers: A_V = 6dB, $R_{L(SP)}$ = 15μH + 8Ω + 15μH, f = 1kHz unless otherwise specified. Limits apply for T_A = 25°C.

Symbol	Parameter	Conditions	LM49270 Typical (2)	LM49270 Limit (3) (4)	Units (Limits)
P_{OUT}	Output Power	Speaker Mode, f = 1kHz,			
		THD+N = 1% R_L = 4Ω R_L = 8Ω	1.75 1.06		W W
		THD+N = 10 % R_L = 4Ω R_L = 8Ω	2.2 1.35		W W
		CC Headphone Mode, f = 1kHz,			
		THD+N = 1% R_L = 16Ω R_L = 32Ω	155 90		mW mW
		THD+N = 10% R_L = 16Ω R_L = 32Ω	177 140		mW mW
		OCL Headphone Mode, f = 1kHz,			
		THD+N = 1% R_L = 16Ω R_L = 32Ω	155 90		mW mW
		THD+N = 10% R_L = 16Ω R_L = 32Ω	175 140		mW mW
THD+N	Total Harmonic Distortion + Noise	Speaker Mode, f = 1kHz P_{OUT} = 100mW, R_L = 8Ω	0.03		%
		CC Headphone Mode, f = 1kHz P_{OUT} = 12mW, R_L = 32Ω	0.02		%
		OCL Headphone Mode, f = 1kHz P_{OUT} = 12mW, R_L = 32Ω	0.03		%
e_N	Noise	Speaker Mode, A-Wtg, Input Referred	47		μV
		CC Headphone Mode, A-Wtg, Input Referred	10		μV
		OCL Headphone Mode, A-Wtg, Input Referred	11		μV
η	Efficiency	Speaker Mode R_L = 8Ω	84		%
Xtalk	Crosstalk	Speaker Mode, f = 1kHz, V_{IN} = 1Vp-p	–85		dB
		CC Headphone Mode, f = 1kHz, V_{IN} = 1Vp-p	–70		dB
		OCL Headphone Mode, f = 1kHz, V_{IN} = 1Vp-p	–58		dB
T_{ON}	Turn-on Time		43		ms
T_{OFF}	Turn-off Time		100		ms
Z_{IN}	Input Impedance	Maximum Gain	23.5		kΩ
		Minimum Gain	210		kΩ
A_V	Gain	Maximum Gain, Speaker Mode	30		dB
		Minimum Gain, Speaker Mode	–47		dB
		Maximum Gain, Headphone Mode	18		dB
		Minimum Gain, Headphone Mode	–59		dB

Electrical Characteristics V_{DD} = 5.0V (continued)

[1] The following specifications apply for Headphone" A_V = 0dB, $R_{L(HP)}$ = 32Ω,: for Loudspeakers: A_V = 6dB, $R_{L(SP)}$ = 15µH + 8Ω + 15µH, f = 1kHz unless otherwise specified. Limits apply for T_A = 25°C.

Symbol	Parameter	Conditions	LM49270		Units (Limits)
			Typical [2]	Limit [3] [4]	
PSRR	Power Supply Rejection Ratio	Speaker Mode, V_{RIPPLE} = 200mVp-p Sine f = 217Hz f = 1kHz	 61 61		 dB dB
		Headphone Mode, V_{RIPPLE} = 200mVp-p Sine, CC Mode f = 217Hz f = 1kHz	 75 74		 dB min
		Headphone Mode, V_{RIPPLE} = 200mVp-p Sine, OCL Mode f = 217Hz f = 1kHz	 78 75		 dB dB
$HPS_{(Th)}$	Headphone Sense Threshold	Detect Headphone		4.4	V (min)
		Detect no Headphone		3	V (max)

Typical Performance Characteristics

THD+N vs Output Power Speaker Mode A_V = 6dB, R_L = 4Ω, f = 1kHz

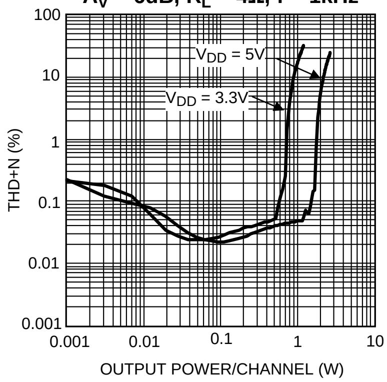

Figure 3.

THD+N vs Output Power Speaker Mode A_V = 6dB, R_L = 8Ω, f = 1kHz

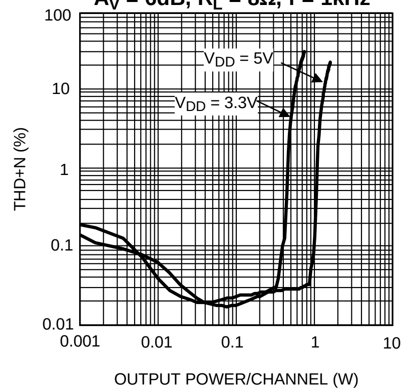

Figure 4.

THD+N vs Output Power OCL Headphone Mode A_V = 0dB, R_L = 16Ω, f = 1kHz

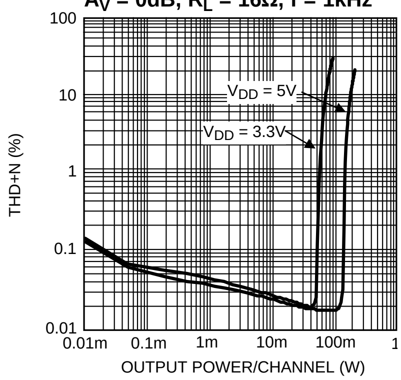

Figure 5.

THD+N vs Output Power OCL Headphone Mode A_V = 0dB, R_L = 32Ω, f = 1kHz

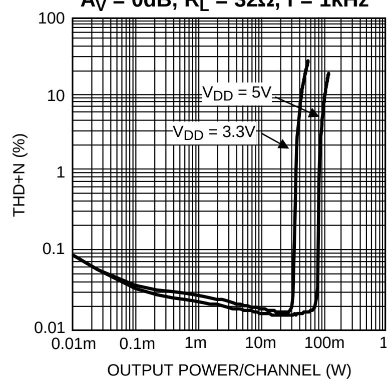

Figure 6.

THD+N vs Output Power CC Headphone Mode A_V = 0dB, R_L = 16Ω, f = 1kHz

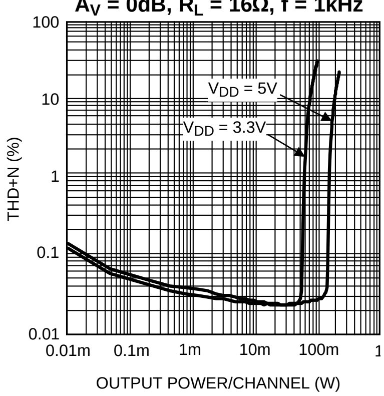

Figure 7.

THD+N vs Output Power CC Headphone Mode A_V = 0dB, R_L = 32Ω, f = 1kHz

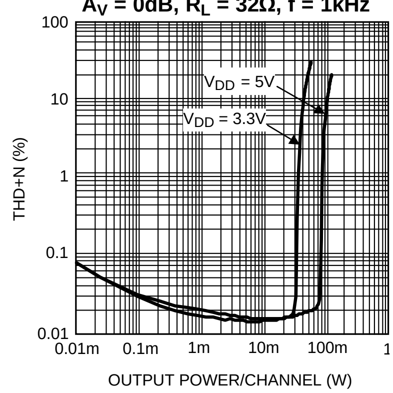

Figure 8.

Typical Performance Characteristics (continued)

THD+N vs Frequency Speaker Mode V_{DD} = 3.3V, P_{OUT} = 300mW, R_L = 4Ω

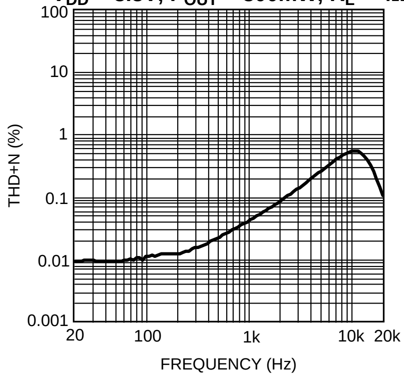

Figure 9.

THD+N vs Frequency Speaker Mode V_{DD} = 5V, P_{OUT} = 500mW, R_L = 4Ω

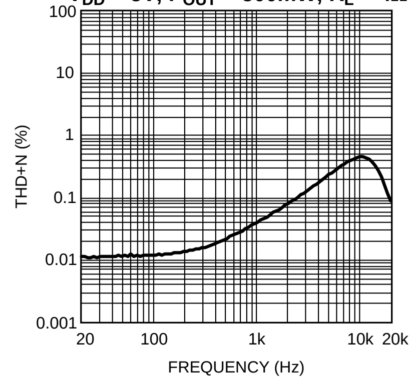

Figure 10.

THD+N vs Frequency Speaker Mode V_{DD} = 3.3V, P_{OUT} = 200mW, R_L = 8Ω

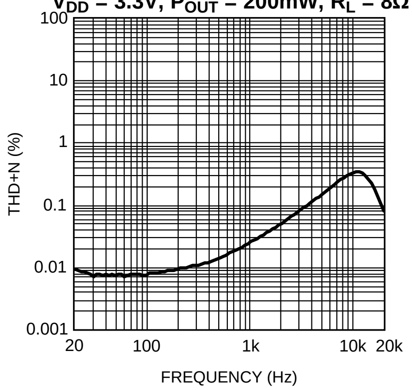

Figure 11.

THD+N vs Frequency Speaker Mode V_{DD} = 5V, P_{OUT} = 350mW, R_L = 8Ω

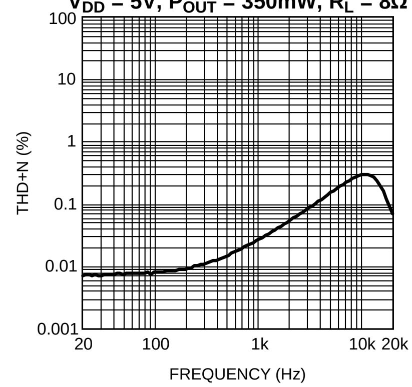

Figure 12.

THD+N vs Frequency OCL Headphone Mode V_{DD} = 3.3V, P_{OUT} = 45mW, R_L = 16Ω

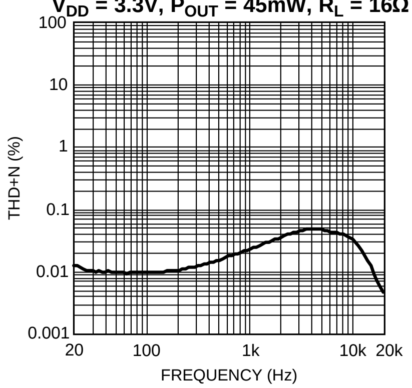

Figure 13.

THD+N vs Frequency OCL Headphone Mode V_{DD} = 5V, P_{OUT} = 100mW, R_L = 16Ω

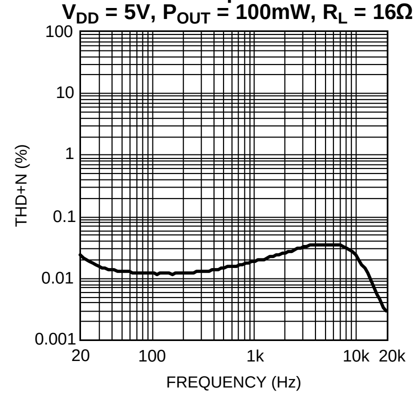

Figure 14.

Typical Performance Characteristics (continued)

THD+N vs Frequency OCL Headphone Mode V_{DD} = 3.3V, P_{OUT} = 25mW, R_L = 32Ω

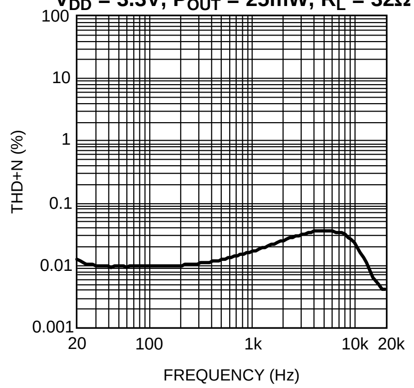

Figure 15.

THD+N vs Frequency OCL Headphone Mode V_{DD} = 5V, P_{OUT} = 70mW, R_L = 32Ω

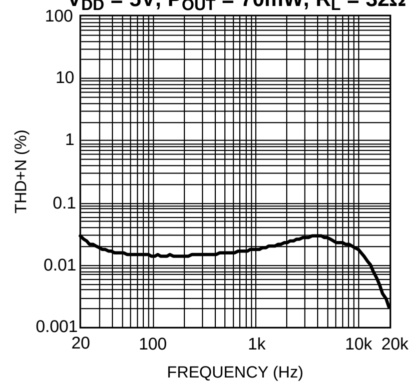

Figure 16.

THD+N vs Frequency CC Headphone Mode V_{DD} = 3.3V, P_{OUT} = 45mW, R_L = 16Ω

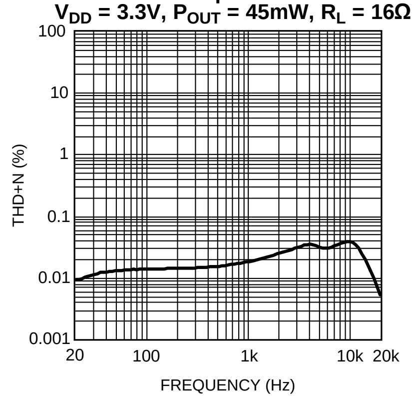

Figure 17.

THD+N vs Frequency CC Headphone Mode V_{DD} = 5V, P_{OUT} = 100mW, R_L = 16Ω

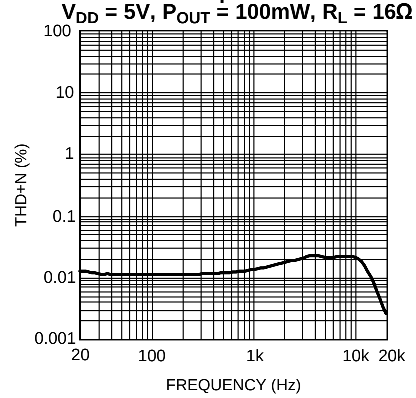

Figure 18.

THD+N vs Frequency CC Headphone Mode V_{DD} = 3.3V, P_{OUT} = 25mW, R_L = 32Ω

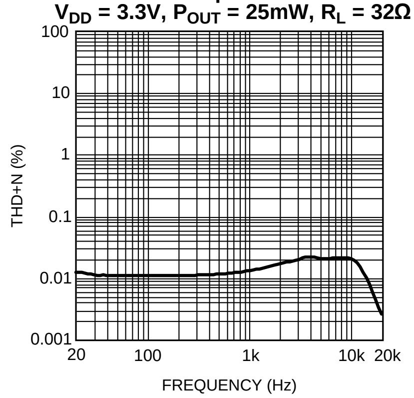

Figure 19.

THD+N vs Frequency CC Headphone Mode V_{DD} = 5V, P_{OUT} = 70mW, R_L = 32Ω

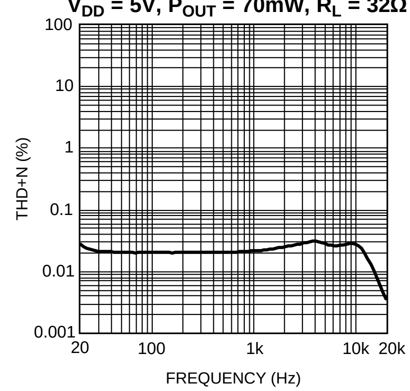

Figure 20.

Typical Performance Characteristics (continued)

PSRR vs Frequency Speaker Mode
V_{DD} = 3.3V, V_{RIPPLE} = 200mV_{P-P}, R_L = 8Ω

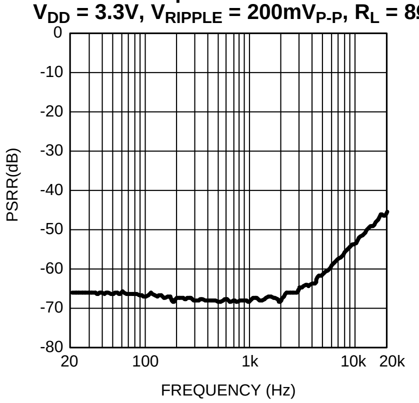

Figure 21.

PSRR vs Frequency OCL Headphone Mode
V_{DD} = 3.3V, V_{RIPPLE} = 200mV_{P-P}, R_L = 32Ω

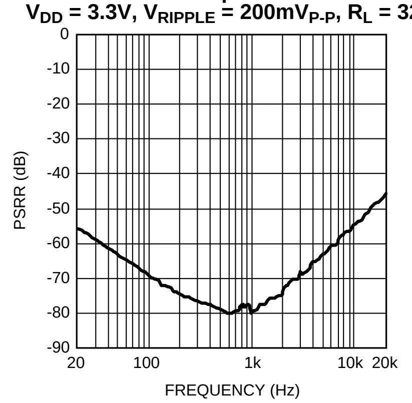

Figure 22.

PSRR vs Frequency CC Headphone Mode
V_{DD} = 3.3V, V_{RIPPLE} = 200mV_{P-P}, R_L = 32Ω

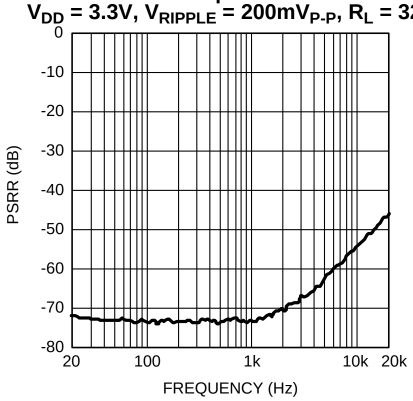

Figure 23.

Efficiency vs Output Power Speaker Mode
R_L = 4Ω, f = 1kHz

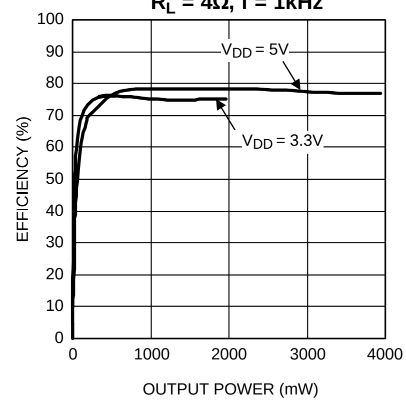

Figure 24.

Efficiency vs Output Power Speaker Mode
R_L = 8Ω, f = 1kHz

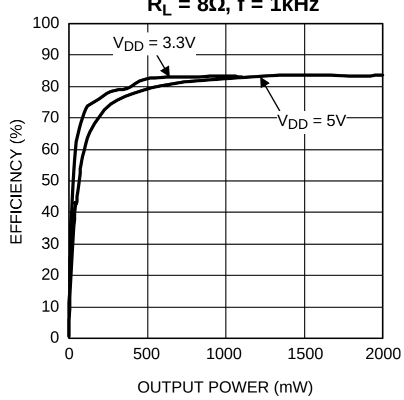

Figure 25.

Power Dissipation vs Output Power Speaker Mode
R_L = 4Ω, f = 1kHz

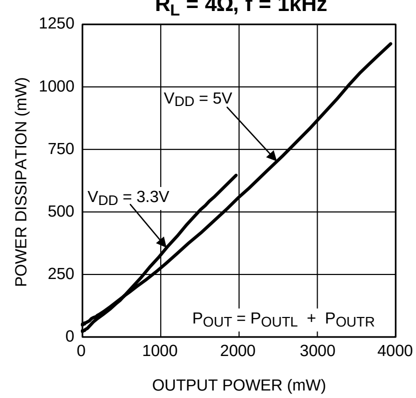

Figure 26.

Typical Performance Characteristics (continued)

Power Dissipation vs Output Power Speaker Mode $R_L = 8\Omega$, f = 1kHz

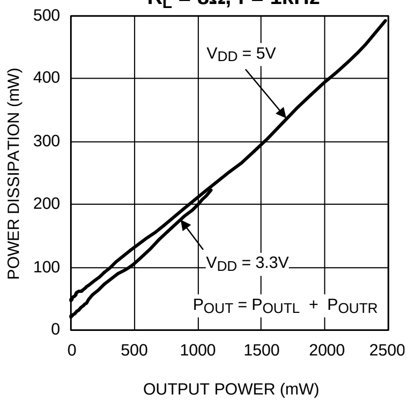

Figure 27.

Power Dissipation vs Output Power OCL Headphone Mode $R_L = 16\Omega$, f = 1kHz

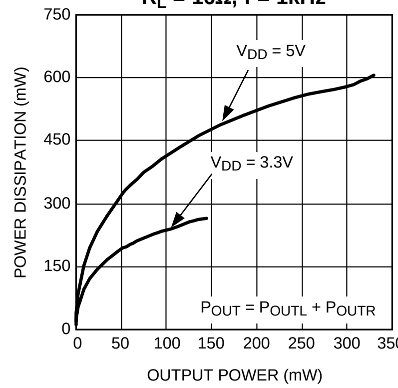

Figure 28.

Power Dissipation vs Output Power OCL Headphone Mode $R_L = 32\Omega$, f = 1kHz

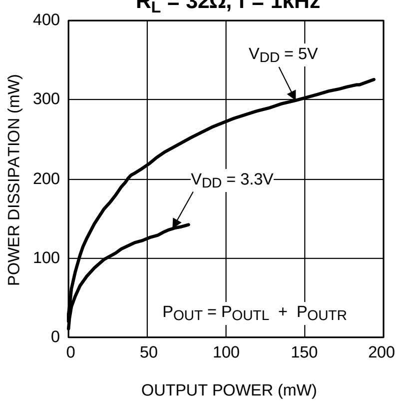

Figure 29.

Power Dissipation vs Output Power CC Headphone Mode $R_L = 16\Omega$, f = 1kHz

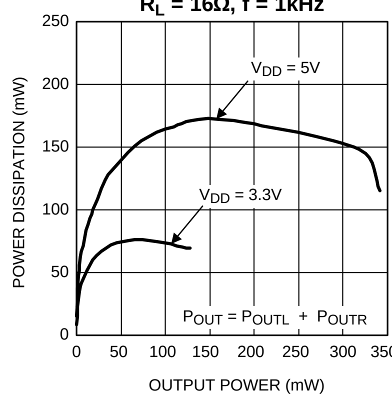

Figure 30.

Power Dissipation vs Output Power CC Headphone Mode $R_L = 32\Omega$, f = 1kHz

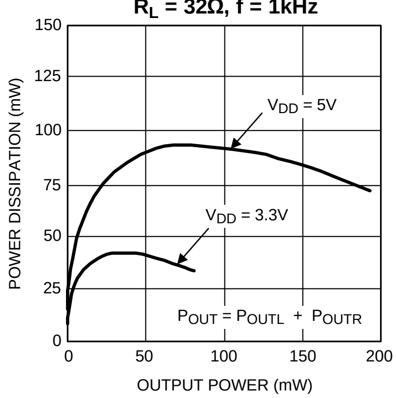

Figure 31.

Output Power vs Supply Voltage Speaker Mode $R_L = 4\Omega$, f = 1kHz

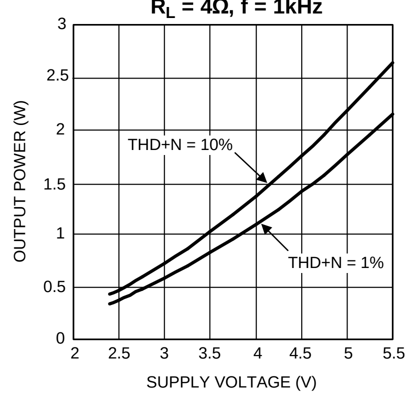

Figure 32.

Typical Performance Characteristics (continued)

Output Power vs Supply Voltage Speaker Mode R_L = 8Ω, f = 1kHz

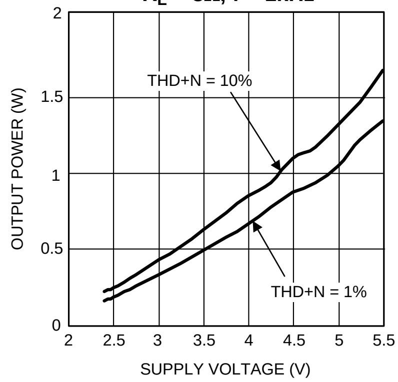

Figure 33.

Output Power vs Supply Voltage OCL Headphone Mode R_L = 16Ω, f = 1kHz

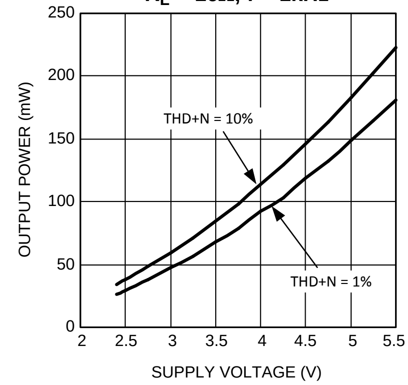

Figure 34.

Output Power vs Supply Voltage OCL Headphone Mode R_L = 32Ω, f = 1kHz

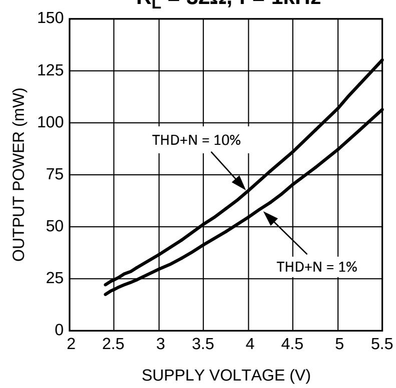

Figure 35.

Output Power vs Supply Voltage CC Headphone Mode R_L = 16Ω, f = 1kHz

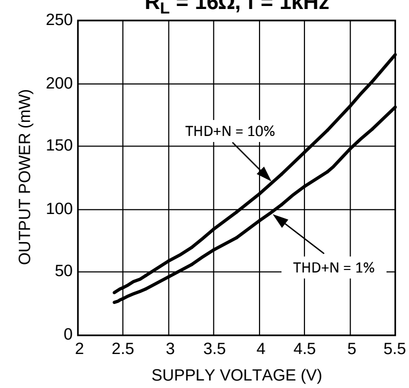

Figure 36.

Output Power vs Supply Voltage CC Headphone Mode R_L = 32Ω, f = 1kHz

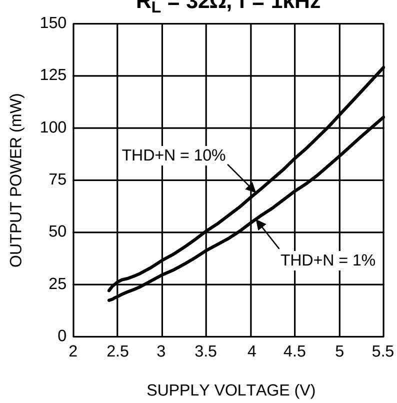

Figure 37.

Crosstalk vs Frequency Speaker Mode V_{DD} = 3.3V, V_{RIPPLE} = $1V_{P-P}$, R_L = 8Ω

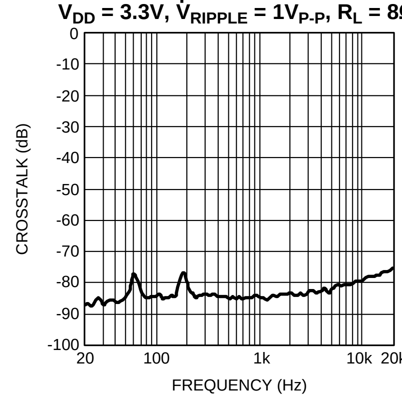

Figure 38.

Typical Performance Characteristics (continued)

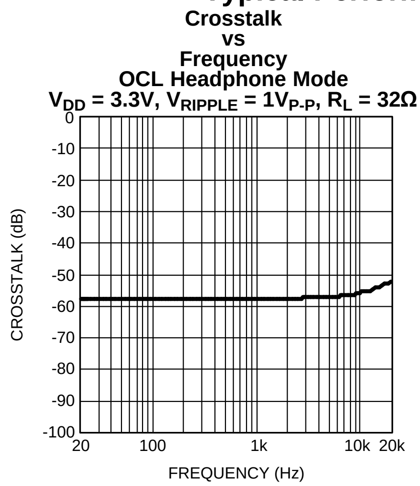

Figure 39.

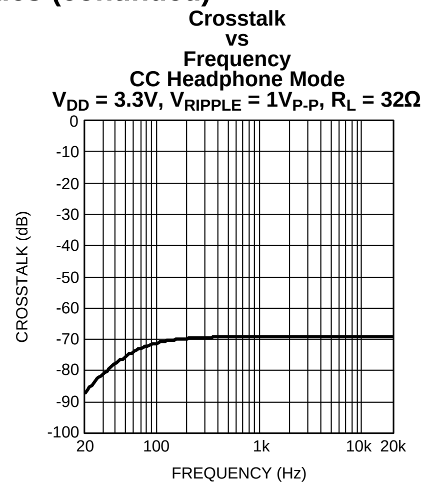

Figure 40.

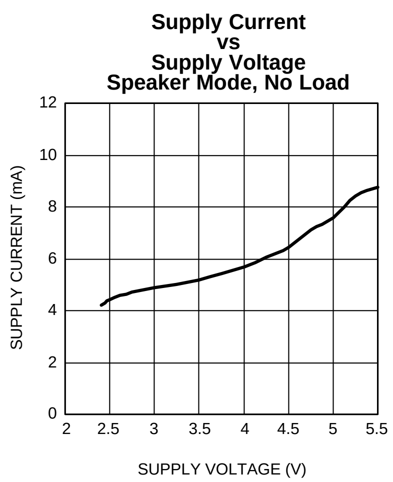

Figure 41.

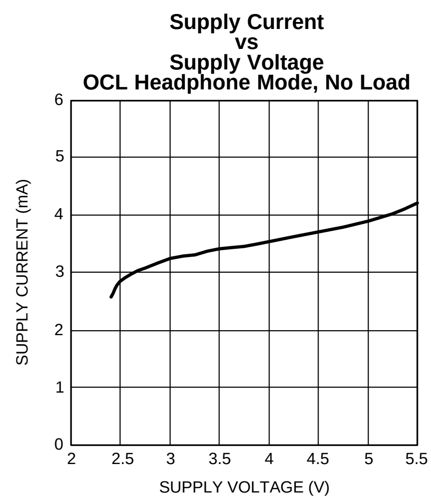

Figure 42.

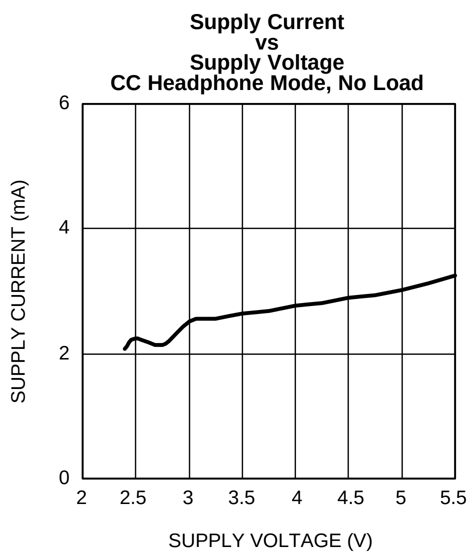

Figure 43.

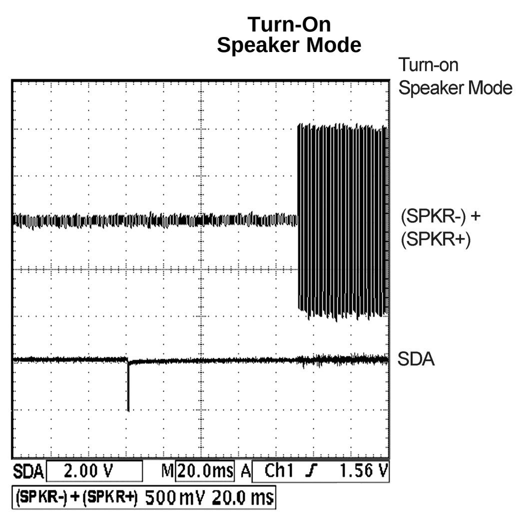

Figure 44.

Typical Performance Characteristics (continued)

Turn-Off
Speaker Mode

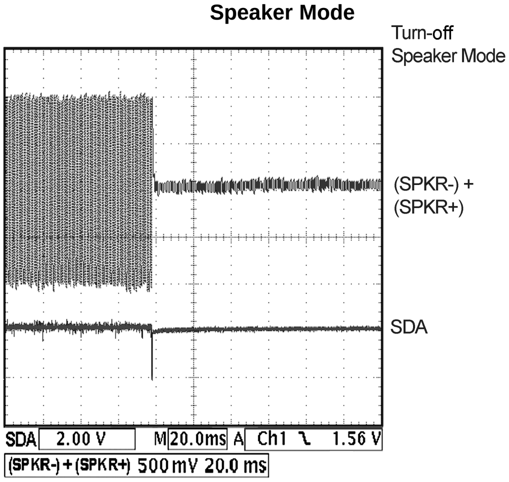

Figure 45.

Turn-On
OCL Headphone Mode

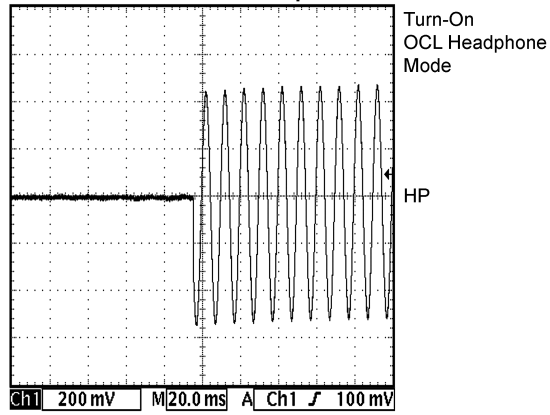

Figure 46.

Turn-Off
OCL Headphone Mode

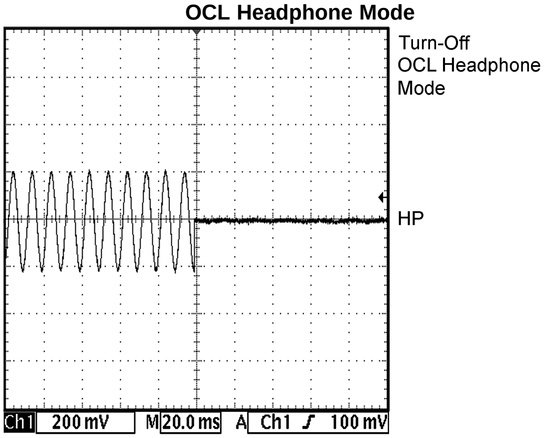

Figure 47.

Turn-On
CC Headphone Mode

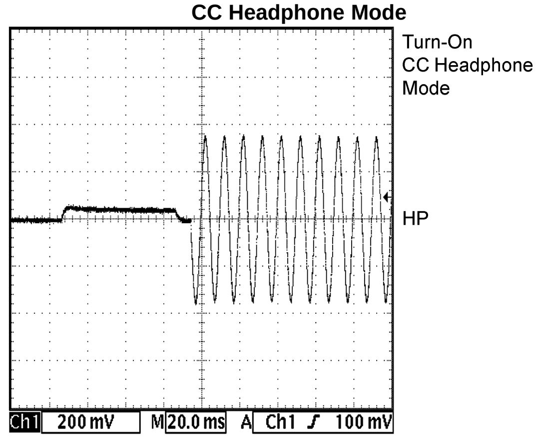

Figure 48.

Turn-Off
CC Headphone Mode

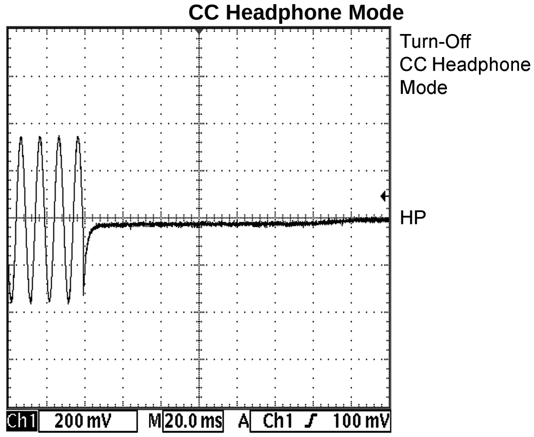

Figure 49.

APPLICATION INFORMATION

I2C COMPATIBLE INTERFACE

The LM49270 is controlled through an I^2C compatible serial interface that consists of a serial data line (SDA) and a serial clock (SCL). The clock line is uni-directional. The data line is bi-directional (open-collector), although the LM49270 does not write to the I^2C bus. The LM49270 and the master can communicate at clock rates up to 400kHz. Figure 51 shows the I^2C interface timing diagram. The LM49270 is a transmit/receive slave-only device, reliant upon the master to generate a clock signal.

The master device communicates to the LM49270 by transmitting the proper device address followed by a command word. Each transmission sequence is framed by a START condition and a STOP condition. Each word (register address + register content) transmitted over the bus is 8 bits long and is always followed by an acknowledge pulse.

To avoid an address conflict with another device on the I^2C bus, the LM49270 address is determined by the ADR pin, the state of ADR determines address bit A1 (Table 1). When ADR = 0, the address is 1111 1000. When ADR = 1 the device address is 1111 1010.

Table 1. Device Address

ADR	A7	A6	A5	A4	A3	A2	A1	A0
X	1	1	1	1	1	0	X	0
0	1	1	1	1	1	0	0	0
1	1	1	1	1	1	0	1	0

Table 2. I^2C Control Registers

REG	Register Name	D7	D6	D5	D4	D3	D2	D1	D0
0	Shutdown Control	0	0	—	—	HP3DSEL	LS3DSEL	OCL/CC	PWR_ON
1	Headphone Gain Control	0	1	—	HP4	HP3	HP2	HP1	HP0
2	Speaker Gain Control	1	0	—	LS4	LS3	LS2	LS1	LS0

NOTE

OCL/CC = 1 selects OCL mode; OCL/CC = 0 selects cap coupled mode

PWR_ON = 0 puts part in shutdown

BUS FORMAT

The I^2C bus format is shown in Figure 50. The “start” signal is generated by lowering the data signal while the clock is high. The start signal alerts all devices on the bus that a device address is being written to the bus.

The 8-bit device address is written to the bus next, most significant bit first. The data is latched in on the rising edge of the clock. Each address bit must be stable while the clock is high.

After the last address bit is sent, the master device releases the data line, during which time, an acknowledge clock pulse is generated. If the LM49270 receives the address correctly, then the LM49270 pulls the data line low, generating an acknowledge bit (ACK).

Once the master device has registered the ACK bit, the 8-bit register address/data word is sent. Each data bit should be stable while the clock level is high. After the 8–bit word is sent, the LM49270 sends another ACK bit. Following the acknowledgement of the data word, the master device issues a “stop” bit, allowing SDA to go high while the clock signal is high.

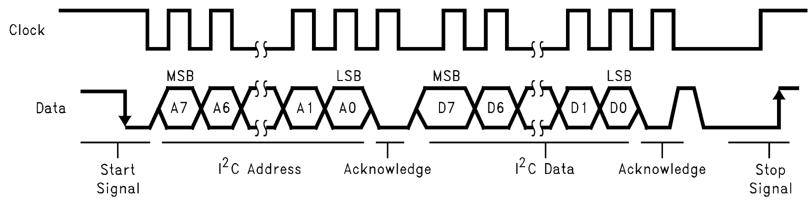

Figure 50. I^2C Bus Format

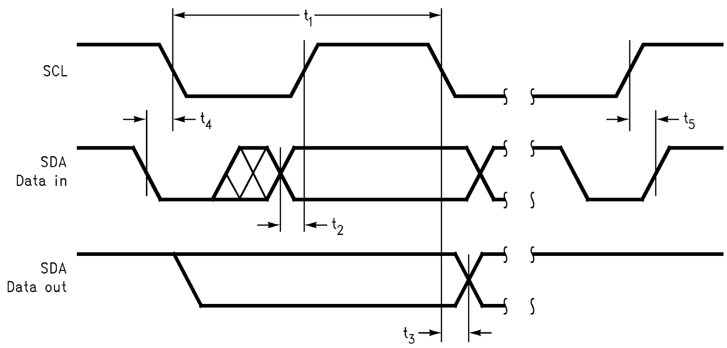

Figure 51. I^2C Timing Diagram

GENERAL AMPLIFIER FUNCTION

Class D Amplifier

The LM49270 features a high-efficiency, filterless, Class D stereo amplifier. The LM49270 Class D amplifiers feature a filterless modulation scheme known as Class BD. The differential outputs of each channel switch at 300kHz from V_{DD} to GND. When there is no input signal applied, the two outputs (LLS+ and LLS-) switch in phase with a 50% duty cycle. Because the outputs of the LM49270 are differential, there is in no net voltage across the speaker, thus no load current during the idle state conserving power.

When an input signal is applied, the duty cycle (pulse width) of each output changes. For increasing output voltages, the duty cycle of LLS+ increases, while the duty cycle of LLS- decreases. For decreasing output voltages, the converse occurs. The duty cycle of LLS- increases while the duty cycle of LLS+ decreases. The difference between the two pulse widths yields the differential output voltage.

Headphone Amplifier

The LM49270 headphone amplifier features two different operating modes, output capacitor-less (OCL) and capacitor coupled (CC). The OCL architecture eliminates the bulky, expensive output coupling capacitors required by traditional headphone amplifiers. The LM49270 headphone section uses three amplifiers. Two amplifiers drive the headphones while the third (VOC) is set to the internally generated bias voltage (typically $V_{DD}/2$). The third amplifier is connected to the return terminal (sleeve) of the headphone jack. In this configuration, the signal side of the headphones are biased to $V_{DD}/2$, the return is biased to $V_{DD}/2$, thus there is no net DC voltage across the headphone eliminating the need for an output coupling capacitor. Removing the output coupling capacitors from the headphone signal path reduces component count, reducing system cost and board space consumption, as well as improving low frequency performance and sound quality. The voltage on the return sleeve is not an issue when driving headphones. However, if the headphone output is used as a line out, the $V_{DD}/2$ can conflict with the GND potential that a line-in would expect on the return sleeve. When the return of the headphone jack is connected to GND, the LM49270 detects an output short circuit condition and the VOC amplifier is disabled preventing damage to the LM49270 and allowing the headphone return to be biased at GND.

Capacitor Coupled Headphone Mode

In capacitor coupled (CC) mode, the VOC pin is disabled, and the headphone outputs are coupled to the jack through series capacitors, allowing the headphone return to be connected to GND (Figure 52). In CC mode, the LM49270 requires output coupling capacitors to block the DC component of the amplifier output, preventing DC current from flowing to the load. The output capacitor and speaker impedance form a high pass filter with a -3dB roll-off determined by:

$$f_{-3dB} = 1 / 2\pi R_L C_{OUT}$$

Where R_L is the headphone impedance, and C_{OUT} is the output coupling capacitor. Choose C_{OUT} such that f_{-3dB} is well below the lowest frequency of interest. Setting f_{-3dB} too high results in poor low frequency performance. Select capacitor dielectric types with low ESR to minimize signal loss due to capacitor series resistance and maximize power transfer to the load.

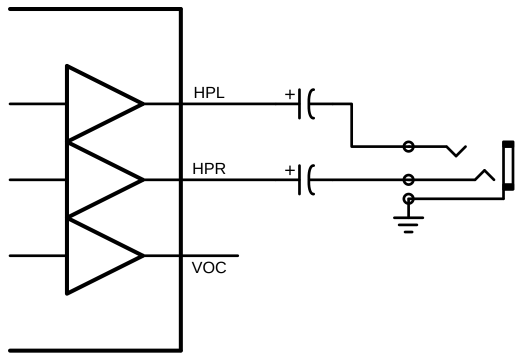

Figure 52. Capacitor Coupled Headphone Mode

Headphone Sense

The LM49270 features a headphone sense input (HPS) that monitors the headphone jack and configures the device depending on the presence of a headphone. When the HPS pin is low, indicating that a headphone is not present, the LM49270 speaker amplifiers are active and the headphone amplifiers are disabled. When the HPS pin is high, indicating that a headphone is present, the headphone amplifiers are active while the speaker amplifiers are disabled.

POWER DISSIPATION AND EFFICIENCY

The major benefit of Class D amplifier is increased efficiency versus Class AB. The efficiency of the LM49270 speaker amplifiers is attributed to the output transistors' region of operation. The Class D output stage acts as current steering switches, consuming negligible amounts of power compared to their Class AB counterparts. Most of the power loss associated with the output stage is due to the IR loss of the MOSFET on-resistance ($R_{DS(ON)}$) , along with the switching losses due to gate charge.

The maximum power dissipation per headphone channel in Capacitor Coupled mode is given by:

$$P_{DMAX(CC)} = V_{DD}^2/2\pi^2 R_L$$

In OCL mode, the maximum power dissipation increases due to the use of a third amplifier as a buffer. The power dissipation is given by:

$$P_{DMAX(OCL)} = V_{DD}^2/\pi^2 R_L$$

SHUTDOWN FUNCTION

The LM49270 features a shutdown mode configured through the I^2C interface. Bit D0 (PWR_ON) in the Shutdown Control register shuts down/turns on the entire device. Set PWR_ON = 1 to enable the LM49270, set PWR_ON = 0 to disable the device.

AUDIO AMPLIFIER GAIN SETTING

Each channel of the LM49270 features a 32 step volume control. The loudspeaker volume has a range of -47dB to 30dB and the headphone has a range of -59dB to 18dB (see Table 3).

Table 3. Volume Control

Volume Step	LS4/HP4	LS3/HP3	LS2/HP2	LS1/HP1	LS0/HP0	LS Gain (dB)	HP Gain (dB)
1	0	0	0	0	0	–47	–59
2	0	0	0	0	1	–36	–48
3	0	0	0	1	0	–28.5	–46.5
4	0	0	0	1	1	–22.5	–34.5
5	0	0	1	0	0	–18	–30
6	0	0	1	0	1	–15	–27
7	0	0	1	1	0	–12	–24
8	0	0	1	1	1	–9	–21
9	0	1	0	0	0	–6	–18
10	0	1	0	0	1	–3	–15
11	0	1	0	1	0	–1.5	–13.5
12	0	1	0	1	1	0	–12
13	0	1	1	0	0	1.5	–10.5
14	0	1	1	0	1	3	–9
15	0	1	1	1	0	4.5	–7.5
16	0	1	1	1	1	6	–6
17	1	0	0	0	0	7.5	–4.5
18	1	0	0	0	1	9	–3
19	1	0	0	1	0	10.5	–1.5
20	1	0	0	1	1	12	0
21	1	0	1	0	0	13.5	1.5
22	1	0	1	0	1	15	3
23	1	0	1	1	0	16.5	4.5
24	1	0	1	1	1	18	6
25	1	1	0	0	0	19.5	7.5
26	1	1	0	0	1	21	9
27	1	1	0	1	0	22.5	10.5
28	1	1	0	1	1	24	12
29	1	1	1	0	0	25.5	13.5
30	1	1	1	0	1	27	15
31	1	1	1	1	0	28.5	16.5
32	1	1	1	1	1	30	18

3D ENHANCEMENT

The LM49720 features TI’s 3D sound enhancement. 3D sound improves the apparent stereo channel separation whenever the left and right speakers are located close to each other, widening the perceived sound stage in devices with a small form factor that prohibits proper speaker placement.

An external RC network , shown in Figure 1, enables the 3D effect. R3D sets the level of the 3D effect; decreasing the value of R3D will increase the 3D effect. The 3D network acts like a high pass filter C3D sets the frequency response; increasing the value of C3D will decrease the low cutoff frequency at which the 3D effect starts to occur, as shown by this equation:

$$f_{3D(-3dB)} = 1/2\pi(R3D)(C3D) \quad (1)$$

Enabling the 3D effect increases the gain by a multiplication factor of (1 + 20kΩ/R3D). Setting R3D to 20kΩ results in a 6dB increase (doubling) of the gain, increasing the 3D effect. The level of 3D effect is also dependent on other factors such as speaker placement and the distance from the speakers to the listener. The values of R3D and C3D should be chosen for each application individually, taking into account the physical factors noted before.

POWER SUPPLIES

The LM49270 uses different supplies for each portion of the device, allowing for the optimum combination of headroom, power dissipation and noise immunity. The speaker amplifier gain stage is powered from V_{DD}, while the output stage is powered from LSV_{DD}. The headphone amplifiers, input amplifiers and volume control stages are powered from HPV_{DD}. The separate power supplies allow the speakers to operate from a higher voltage for maximum headroom, while the headphones operate from a lower voltage, improving power dissipation. HPV_{DD} may be driven by a linear regulator to further improve performance in noisy environments. The I^2C portion if powered from I^2CV_{DD}, allowing the I^2C portion of the LM49270 to interface with lower voltage digital controllers.

PROPER SELECTION OF EXTERNAL COMPONENTS

Audio Amplifier Power Supply Bypassing/Filtering

Proper power supply bypassing is critical for low noise performance and high PSRR. Place the supply bypass capacitor as close to the device as possible. Typical applications employ a voltage regulator with 10µF and 0.1µF bypass capacitors that increase supply stability. These capacitors do not eliminate the need for bypassing of the LM49270 supply pins. A 1µF capacitor is recommended.

Bypass Capacitor Selection

The LM49270 generates a $V_{DD}/2$ common-mode bias voltage internally. The BYPASS capacitor, C_B, improves PSRR and THD+N by reducing noise at the BYPASS node. Use a 1µF capacitor, placed as close to the device as possible for C_B.

Audio Amplifier Input Capacitor Selection

Input capacitors, C_{IN}, in conjunction with the input impedance of the LM49270 forms a high pass filter that removes the DC bias from an incoming signal. The AC-coupling capacitor allows the amplifier to bias the signal to an optimal DC level. Assuming zero source impedance, the -3dB point of the high pass filter is given by:

$$f_{(-3dB)} = 1/2\pi R_{IN} C_{IN} \quad (2)$$

Choose C_{IN} such that f_{-3dB} is well below that lowest frequency of interest. Setting f_{-3dB} too high affects the low-frequency responses of the amplifier. Use capacitors with low voltage coefficient dielectrics, such as tantalum or aluminum electrolytic. Capacitors with high-voltage coefficients, such as ceramics, may result in increased distortion at low frequencies. Other factors to consider when designing the input filter include the constraints of the overall system. Although high fidelity audio requires a flat frequency response between 20Hz and 20kHz, portable devices such as cell phones may only concentrate on the frequency range of the frequency range of the spoken human voice (typically 300Hz to 4kHz). In addition, the physical size of the speakers used in such portable devices limits the low frequency response; in this case, frequencies below 150Hz may be filtered out.

REVISION TABLE

Rev	Date	Description
1.0	12/19/06	Initial release.

PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
LM49270SQ/NOPB	ACTIVE	WQFN	RSG	28	1000	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 85	49270SQ	Samples
LM49270SQE/NOPB	ACTIVE	WQFN	RSG	28	250	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 85	49270SQ	Samples

(1) The marketing status values are defined as follows:
ACTIVE: Product device recommended for new designs.
LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.
NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.
PREVIEW: Device has been announced but is not in production. Samples may or may not be available.
OBSOLETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.
TBD: The Pb-Free/Green conversion plan has not been defined.
Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.
Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.
Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

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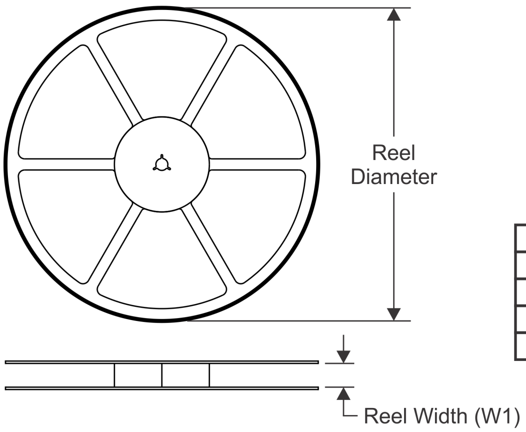
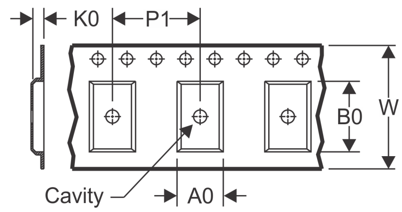
TAPE AND REEL INFORMATION

REEL DIMENSIONS

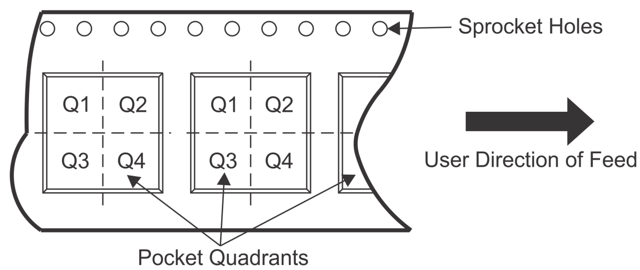


TAPE DIMENSIONS



A0	Dimension designed to accommodate the component width
B0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

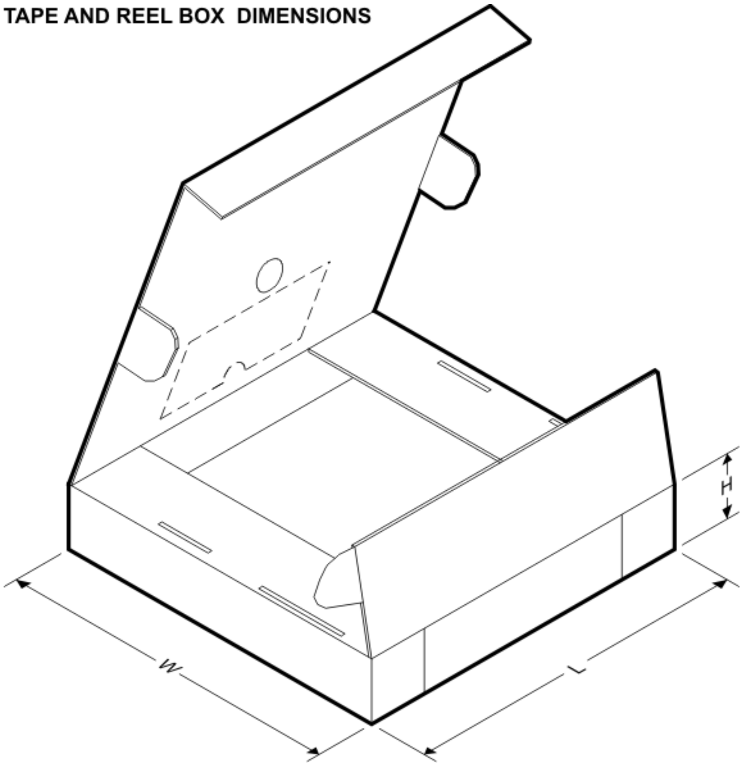
QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
LM49270SQ/NOPB	WQFN	RSG	28	1000	178.0	12.4	5.3	5.3	1.3	8.0	12.0	Q1
LM49270SQE/NOPB	WQFN	RSG	28	250	178.0	12.4	5.3	5.3	1.3	8.0	12.0	Q1

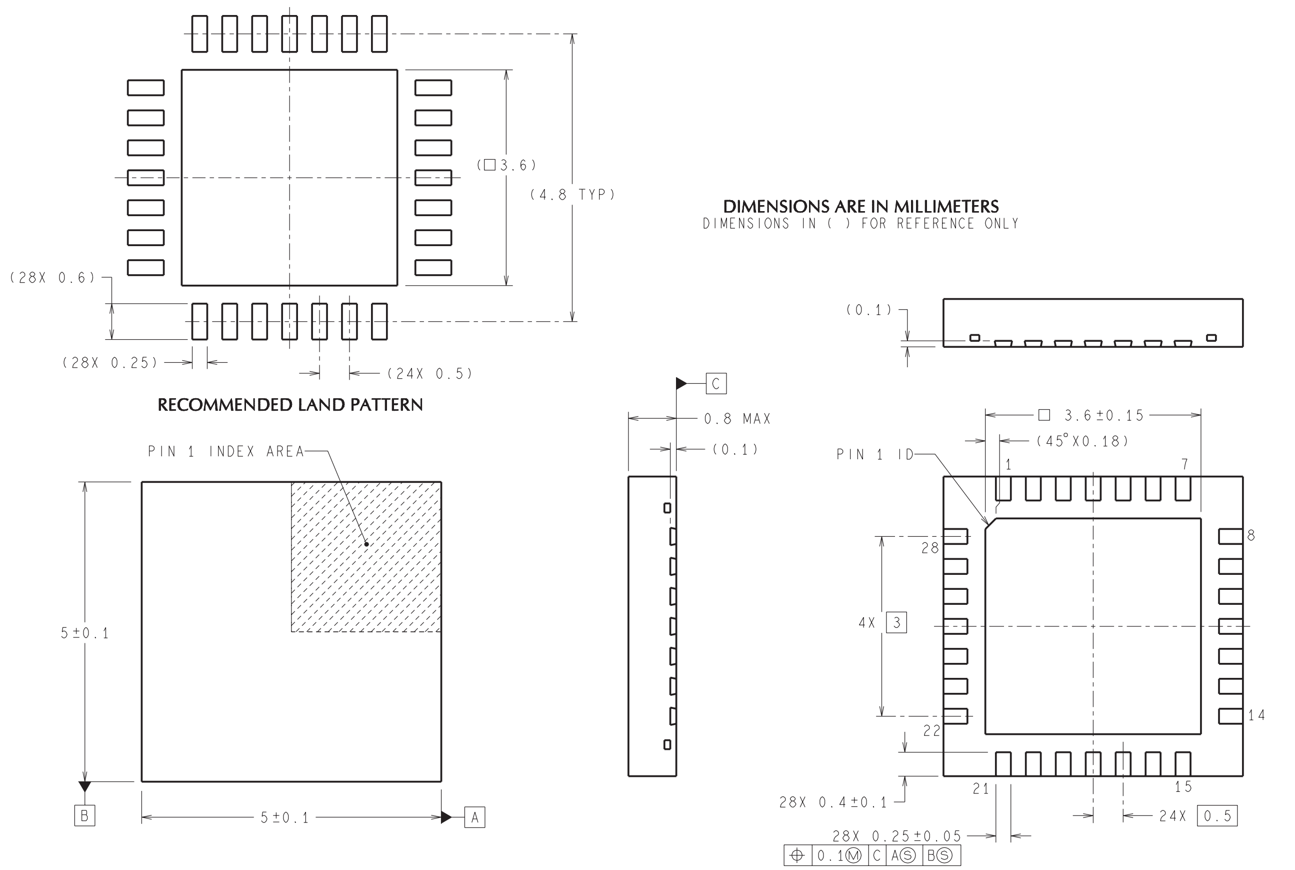
TAPE AND REEL BOX DIMENSIONS



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
LM49270SQ/NOPB	WQFN	RSG	28	1000	210.0	185.0	35.0
LM49270SQE/NOPB	WQFN	RSG	28	250	210.0	185.0	35.0

RSG0028A



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